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Solution of Large Transmission Line Type Circuits Using a New Optimized Waveform Relaxation Partitioning

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Abstract

Larger problems can be solved with the ever increasing availability of computing power. However, our ability to solve large circuit and transmission line problems, in the time domain, is still relatively modest. Partitioning into a set of smaller problems is essential for the electrical modeling of very large systems. However, so far the partitioning of strongly coupled circuits like transmission lines has been difficult. We introduce a new approach for coupling partitioned problems in the context of waveform relaxation methods, which leads to the class of optimized waveform relaxation algorithms with greatly enhanced performance. We analyze convergence for the case of a transmission line type circuit and illustrate the method with several quasi TEM mode transmission line problems.

1 Introduction

Presently, the performance of electronic systems is increasing rapidly as the frequencies and cycle times are getting higher and as the rise time of the signals is decreasing. This leads to numerous new challenges for the Electrical Interconnect and Package (EIP) and the EMC electromagnetic modeling of realistic structures. Probably the most challenging aspect, which results from the faster signal rise times and higher frequencies, is the increase in the system or circuit size. It is clear that conventional solution methods will slowly make progress toward larger systems mostly due to the increase in available computing power. However, since the size of a system is limited, it is very important to be able to partition large problems into smaller ones without loss of accuracy.

In general, time domain methods have a great advantage over spectral methods from a partitioning point of view since a natural decoupling occurs due to the spatial time retardation. Unfortunately, coupling extends over the entire solution space in the frequency domain. From an EM solver point of view, we see a gradual shift of the environment and interfaces from purely electromagnetic toward Spice circuit solver interfaces which process circuits as well as the electromagnetic parts. Hence, the solver approaches must be able to handle EM parts as well as Spice type circuits.

Partitioning is essential for the electrical modeling of very large systems. In this paper we introduce a new approach for the coupling of partitioned systems. We illustrate the method with a quasi TEM mode transmission line model problem and then show numerical experiments for several larger test cases. The underlying technique, which makes use of the time domain behavior and its decoupling aspects is called waveform relaxation (WR). It has its origin in the Spice type circuit solver area [1] and it has been applied to many different problems [2]. In this approach, large circuits are subdivided into sub-circuits which are then solved individually, usually for a window in time while the time axis is divided into multiple windows. A relaxation procedure is then used to iterate among the sub-circuits until convergence is achieved. A summary of the application of the classical WR to circuit problems is given in [3]. The new optimized WR approach was first ap-
plied to partial differential equation problems in [4]. We recognized that the new optimized WR method is very much suitable for all techniques which can be formulated in the circuit domain, see for example [5] for RC type circuits. One of the well known aspects of the conventional, classical WR is that the partitioning of the models has to be done very carefully to avoid an excessive number of WR iterations. This will also be evident from the work in this paper. It is very desirable to keep the number of WR iterations low, even for the largest problems to be solved. Usually, the relative error in time domain EM and Spice type circuit solvers is $10^{-2}$ to $10^{-3}$. We want to note that even the classical WR approach is well suited for parallel computations since it fits into the class of *highly parallelizable* methods. This has been substantiated by partitioning a very large 180k transistor circuit into about $10^3$ sub-circuits or subsystems. An average speed up of about 180 was achieved on a system with 265 processors [6]. Hence, the suitability for parallel processing of the new WR algorithm is an aspect which we strongly emphasize.

In this paper we show how the new optimized WR technique can be applied to the partitioning of quasi TEM mode transmission line models. We derive the new algorithm for a small model problem in Section 2. In Section 3 we then show that the technique derived for the small model problem also greatly enhances the performance of the algorithm when it is applied to large transmission line circuits also with variable elements. We consider here the improvement for the new optimized WR approach over the classical WR for the efficient solution of partitioning transmission lines.

2 The Transmission Line Model

First, the small quasi TEM mode transmission line example given in Figure 1 is analyzed with the new approach. To show the basic idea of the WR partitioning, we divide the example into two sub-circuits which is equivalent to two sections of a lumped quasi TM transmission line model. This corresponds to breaking the line into two parts. We analyze the model using the classical as well as optimized WR algorithm to compare the convergence behavior for the small model lumped circuit. The circuit equations are given in the form of (2.1) which is convenient for the convergence analysis,

$$\dot{x} = \begin{bmatrix} b_1 & c_1 & \dots & \end{bmatrix} x + f. \quad (2.1)$$

The vector of unknown waveforms $x = (v_1, i_1, v_2, i_2, v_3)^T$ which consists of nodal voltages and inductance currents is the same we use in the Modified Nodal Analysis (MNA) technique. The elements of the matrix are given by

$$a_i = \frac{1}{L_{(i+1)/2}}, \quad a_i = \frac{1}{C_{(i+1)/2}}, \quad (i \text{ even}),$$

$$b_1 = \frac{-1}{R_s C_1}, \quad b_i = 0, \quad (i > 1, \text{ odd}),$$

$$b_i = \frac{R_{i/2}}{L_{i/2}}, \quad b_5 = \frac{-1}{R_L C_3},$$

$$c_i = \frac{-1}{C_{(i+1)/2}}, \quad (i \text{ odd}), \quad c_i = \frac{-1}{L_{i/2}}, \quad (i \text{ even}),$$

and the source on the right hand side is given by $f(t) = (I_s/C_1, 0, 0, 0, 0)^T$. We also need the initial values $x(0) = (v_1^0, i_1^0, v_2^0, i_2^0, v_3^0)^T$ to start the transient analysis.

2.1 The Classical WR Algorithm

We partition the circuit into two sub-circuits or subsystems and we call the unknown values in subsystem 1 $u(t)$ while the unknowns in subsystem 2 are called $w(t)$. Then the classical WR algorithm applied to (2.1) with two sub-circuits is given by

$$\dot{u}^{k+1} = \left[ \begin{array}{ccc} b_1 & c_1 & \dots \\ a_1 & b_2 & c_2 \\ & a_2 & b_3 & \dots \end{array} \right] u^{k+1} + \begin{bmatrix} f_1 \\ f_2 \\ f_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ c_3 u_{1}^k \end{bmatrix},$$

$$\dot{w}^{k+1} = \left[ \begin{array}{ccc} b_4 & c_4 & \dots \\ a_4 & b_5 & \dots \end{array} \right] w^{k+1} + \begin{bmatrix} f_4 \\ f_5 \end{bmatrix} + \begin{bmatrix} a_3 u_{3}^k \\ 0 \end{bmatrix},$$

(2.2)
with corresponding initial conditions $u_{k+1}^{0}(0) = (v_{1}^{0}, i_{1}^{0}, v_{2}^{0})^T$ and $w_{k+1}^{0}(0) = (v_{2}^{0}, v_{3}^{0})^T$. To start
the WR iteration, we need to specify five initial waveforms $u^{0}(t) = (u_{1}^{0}(t), u_{2}^{0}(t), u_{3}^{0}(t))^T$ and $w^{0}(t) = (w_{1}^{0}(t), w_{2}^{0}(t))^T$ for $t \in [0, T]$ where $T$ is
the end of the transient analysis interval. It is well
known that the classical WR algorithm does not work well for transmission line problems, convergence is very slow and initially the algorithm often
seems to diverge, as we will also see in the numerical
experiments in Section 3. We introduce in the
next Subsection a remedy for this problem.

2.2 The Optimized WR Algorithm

The key improvements in our new WR algorithm are better transmission conditions than the ones
used for the classical WR algorithm. It is evident
that the WR convergence will be slow if the in-
formation exchange at the interface between two
subsystems is ineffective. We first emphasize that
the classical transmission conditions employed in
(2.2) can be written explicitly as
\[ u_{4}^{k+1} = w_{1}^{k}, \quad w_{0}^{k+1} = u_{3}^{k}. \tag{2.3} \]

From this we see that in the first sub-circuit the
current $u_4$ is directly replaced in (2.2) by a cur-
rent source, whereas in the second sub-circuit
the voltage $w_0$ is directly replaced by a voltage
source. Hence sub-circuit 1 passes only voltage
information to sub-circuit 2, while sub-circuit 2
only passes current information to sub-circuit 1.
At convergence we obtain with the classical trans-
mision conditions
\[ u_{4}^{\infty} = w_{1}^{\infty}, \quad w_{0}^{\infty} = u_{3}^{\infty}. \tag{2.4} \]

Under these conditions, the nodes at the subsys-
tem boundaries assume the converged current and
voltage respectively, as expected.

For the optimized WR algorithm we propose trans-
mision conditions which exchange both cur-
tent and voltage information in both directions,
\[ u_{4}^{k+1} + \alpha u_{3}^{k+1} = w_{1}^{k} + \alpha w_{0}^{k}, \]
\[ u_{1}^{k+1} + \beta w_{0}^{k+1} = u_{4}^{k} + \beta w_{3}^{k}. \tag{2.5} \]

As can be ascertained by comparing (2.3) with
the new transmission conditions (2.5) we now ex-
change a combination of voltage and current in
both directions, and we introduced weighting fac-
tors $\alpha$ and $\beta$ which can be used to optimize the
performance of the new waveform relaxation algo-

Note that the new transmission conditions lead to the correct solution of the underlying TEM
circuit equations if the new WR algorithm con-
verges and if $\alpha \neq \beta$, because then at convergence
we have from (2.5)
\[ (u_{4}^{\infty} - w_{1}^{\infty}) + \alpha (u_{3}^{\infty} - w_{0}^{\infty}) = 0, \]
\[ (u_{1}^{\infty} - w_{0}^{\infty}) + \beta (u_{3}^{\infty} - w_{0}^{\infty}) = 0 \]

and the determinant of this system is different from
zero, if $\alpha \neq \beta$. Hence the old transmission
conditions (2.3) are implied by the new ones.

For the optimized WR algorithm, the equivalent
to the classical WR algorithm (2.2) is
\[ \dot{u}_{k+1} = \begin{bmatrix} b_{1} c_{1} \\ a_{1} b_{2} c_{2} \end{bmatrix} u_{k+1} + \begin{bmatrix} f_{1} \\ f_{3} \end{bmatrix} + \begin{bmatrix} 0 \\ c_{3}(w_{1}^{k} + \alpha w_{0}^{k}) \end{bmatrix}, \]
\[ \dot{w}_{k+1} = \begin{bmatrix} b_{4} - \frac{a_{3}}{a_{2}} c_{1} \\ a_{4} b_{5} \end{bmatrix} w_{k+1} + \begin{bmatrix} f_{3} \\ f_{4} \end{bmatrix} + \begin{bmatrix} \frac{a_{3}}{a_{2}} (\beta w_{0}^{k} + u_{3}^{k}) \end{bmatrix}, \tag{2.6} \]

where the values $u_{k}^{k}$ and $w_{0}^{k}$ are determined by the transmission conditions (2.5).

The key question is how to choose $\alpha$ and $\beta$ to ob-
tain a WR algorithm with excellent performance
for transmission line problems. We have devel-
oped a theory which allows us to determine the
optimal parameters, but this theory is beyond the
scope of this short paper. It leads to a small opti-
mization problem to be solved for the optimal pa-
rameters. With these parameters, the algorithm
becomes what we call an optimized waveform relax-
ation algorithm.

3 Numerical Experiments

We first perform numerical experiments on the
small model transmission line given in Figure 1.
We use the circuit parameters $C_{i} = 0.63$ pF,
for $i = 1, 2, 3, R_{i} = 0.5e - 3$ kOhms, $i = 1, 2, 3$, $L_{i} = 4.95e - 3$ $\mu$H, $i = 1, 2$ with $R_{3} = 0.05$ kOhms
and $R_{1} = 0.05$ kOhms. We use the source $I_{s} = 10t$
for $0 < t < 0.1$ and $I_{s} = 1$ mA for $t \geq 0.1$ and
the analysis time interval is $[0, T]$ with $T = 1$ ns. To
illustrate the main difference between the classical
and optimized WR algorithm we first show itera-
tions 2, 4 and 6 for the classical WR and the opti-
mized WR in Figure 2. The much faster and more
uniform convergence of the optimized WR algorithm is evident from this computer experiment. The optimal parameters we used were $\alpha = -5.955$ and $\beta = 14.411$. In Figure 3 we show the error as a function of the WR iterations. The remarkable improvement of the optimized WR algorithm over the classical one is evident from this comparison where the classical WR is shown not to converge in less than 20 iterations in the 1 ns time window, while the optimized WR decreases the error to $10^{-6}$ in the same number of iterations.

Next, we show the analysis of larger circuits. The first experiment is to show that the technique developed for the small model circuit also works for large transmission lines. We chose a transmission line which consists of one hundred elements with the same topology as the small circuit in Figure 1. We partition the circuit into two sub-circuits of equal size and compare again the classical waveform relaxation algorithm to the new, optimized one, which uses the transmission conditions (2.5). We use the same circuit parameters and the same source term as for the small circuit, but for a longer transient analysis time interval $[0, T]$, $T = 5$ ns, to give the signal time to travel along the line. The performance of the classical and optimized waveform relaxation algorithm on this larger circuit is shown in Figure 4. Again the optimized WR algorithm converges very rapidly, while the classical WR does not converge in 20 iterations on the 5 ns time window. It is important to note that we have used here the same optimization parameter values $\alpha = -5.955$ and $\beta = 14.411$ that were found for the small circuit, the optimization is a local one and the resulting
parameters work equally well on the larger circuit.

In the last sequence of numerical experiments we show that the optimized waveform relaxation algorithm also works well if the transmission line components are not uniform, which shows its robustness. We chose a transmission line with eight elements similar to the two used to build the small circuit in Figure 1, partition the circuit again into two circuits of equal size and compare the classical waveform relaxation algorithm to the new, optimized one, which uses the transmission conditions (2.5). As a reference, we first use the same constant circuit parameters as in the earlier experiments and integrate on the time window $[0, T]$ with $T = 2$ ns, which leads to the convergence behavior of the classical and optimized WR shown in Figure 5. We now repeat the same experiment for a circuit with moderately varying parameters (up to 50 %), as shown graphically in Figure 6 on top. Again we use the same values for the optimization parameters $\alpha$ and $\beta$ obtained for the small model transmission line circuit with constant circuit elements. As one can see from the convergence behavior in Figure 6 at the bottom, the classical WR is very much affected by the varying circuit elements, it diverges even more rapidly on the first 2 ns time window than in the constant element case shown in Figure 5, whereas the optimized WR converges to the same tolerance as before for the constant element reference case. We finally performed an experiment for strongly varying circuit parameters for a medium size TL shown graphically on the top, and the performance of classical and optimized WR for this problem below.
Figure 7: Large variation in the circuit parameters for a medium size TL shown graphically on top, and performance of the classical and optimized WR for this problem below.

We have introduced an optimized WR algorithm for the partitioned time domain analysis of transmission line problems. The approach uses new transmission conditions which are adapted to the circuit and exchange both current and voltage information. We have shown numerical experiments that illustrate the large improvement in the convergence behavior of the new algorithm over classical WR. In fact, it would not be practical to use the classical WR for the important strongly connected structure of a transmission line. We also showed that the convergence rate of the optimized WR is rather insensitive for a nonuniform transmission line.

4 Conclusions

We have introduced an optimized WR algorithm for the partitioned time domain analysis of transmission line problems. The approach uses new

References


